# TECHNICAL TRANSACTIONS 4/2018

## **ELECTRICAL ENGINEERING**

DOI: 10.4467/2353737XCT.18.059.8371 SUBMISSION OF THE FINAL VERSION: 28/3/2018

## Leonid Moroz

Department of Security of Information Technology, Lviv Polytechnic National University, Ukraine

## Volodymyr Samotyy (vsamotyy@pk.edu.pl)

Department of Automatic Control and Information Technology, Cracow University of Technology

# The CORDIC method of calculating the exponential function Metoda CORDIC obliczania funkcji eksponencjalnej

#### Abstract

This article presents a modern method of calculating the exponential function exp(x) based on the CORDIC iterative algorithm. The proposed solution is implemented in the form of a single iterative equation, which results in the simplification of the electronic version of this algorithm, thus reducing the cost of the device. It is important to point out that the accuracy of the calculation of the analysed function is not lost. **Keywords:** exponential function, hybrid CORDIC method, LUT lookup table, iterative equations, activation functions

#### Streszczenie

W artykule zaproponowano nowoczesną metodę obliczania funkcji eksponencjalnej exp(x) na bazie algorytmu iteracyjnego CORDIC. Przedstawione rozwiązanie jest realizowane w postaci jednego równania iteracyjnego, co prowadzi do uproszczenia wersji elektronicznej tego algorytmu i w wyniku zmniejsza koszt urządzenia. Należy podkreślić, że przy tym nie tracimy na dokładności obliczania badanej funkcji.

Słowa kłuczowe: funkcja eksponencjalna, hybrydowa metoda CORDIC, tablica odnośników LUT, równania iteracyjne, funkcje aktywacji

## 1. Introduction

CORDIC (coordinate rotation in a digital compute) algorithms were developed a long time ago; however, to this day they remain a subject of interest to many researchers. The method has been developed by Jack E. Volder [9]. The method can be easily used for fast Fourier transform calculations, Householder transformation calculations, digital signal filtering, image recognition, etc. The idea of this algorithm involves the application of the iterative process of rotating vectors on a plane at any angle – only shifting and adding operations are used. A number of patents has been published in which CORDIC algorithms are implemented in the form of electronic devices; however, only one of them features an electronic implementation of the exponential function [10]. This approach involves a large amount of computation and requires the electronic implementation of two iterative equations, which is a time-consuming process. There is a well-known paper on simplified exponential function calculation in which only two iterations are used [7, pp. 1–4]. This simplifies the electronic implementation of the CORDIC algorithm, although in the proposed approach, the number of calculations remains unchanged.

## 2. Calculating the exponential function using the hybrid CORDIC method

The device designed for calculating functions is

$$x = \exp(\pm \varphi) , \qquad (1)$$

Measurements of the input signal  $\varphi$  are converted into binary code (*m*– number of bits)

$$\varphi = \sum_{i=1}^{m} a_i 2^{-i}, \ a_i = \{0,1\}, \ \varphi \in [0,1]$$
(2)

The following well-known exponential function calculators [1-6, 8] use CORDIC iterative methods (Fig. 1):

$$x_{i+1} = x_i + \sigma_i y_i 2^{-i}, (3)$$

$$y_{i+1} = y_i + \sigma_i x_i 2^{-i}, (4)$$

$$z_{i+1} = z_i - \sigma_i \alpha_i, \qquad (5)$$

$$\alpha_i = \arctan h(2^{-i}), \ i = 1, 2, 3, 4, 4, 5 \dots 12, 13, 13, 14, \dots, m$$
, (6)

$$\sigma_i = \begin{cases} -1 \text{ if } z_i < 0\\ +1 \text{ if } z_i \ge 0 \end{cases}, \tag{7}$$





Fig. 1. Electronic implementation of the CORDIC method

$$x_1 = P', y_1 = 0, z_1 = \phi, x_{m+1} \approx \cosh(\phi), y_{m+1} \approx \sinh(\phi), \phi \in [0, 1.118]$$
 (8)

$$P_{m1}^{'} = \prod_{i=1}^{m} \cosh(\alpha_{i}) = \prod_{i=1}^{m} \frac{1}{\sqrt{1 - 2^{-2i}}}; P_{m2}^{'} = \frac{1}{\sqrt{1 - 2^{-8}}} \frac{1}{\sqrt{1 - 2^{-26}}} \frac{1}{\sqrt{1 - 2^{-80}}}$$
(9)

$$P' = P_{m1}' P_{m2}'$$
(10)

After the completion of the iterative calculation of the function  $y_{m+1} \approx \sinh(\varphi)$  and  $x_{m+1} \approx \cosh(\varphi)$ , we can calculate the exponent functions:

$$x_{m+1} + y_{m+1} \approx \exp(\phi), \ x_{m+1} - y_{m+1} \approx \exp(-\phi)$$
 (11)

The disadvantage of such devices is the redundancy of electronic components for the implementation of three iterative equations for variables  $x_{i+1}$ ,  $y_{i+1}$  and  $z_{i+1}$ , and the resulting long calculation time (for a calculation with accuracy of *m* bits, we have to implement *m*+2 iterative cycles). The simplified device (Fig. 2) for the exponent calculation is described in [7]. The device implements the algorithm

$$w_{i+1} = w_i + \sigma_i w_i 2^{-i}, \ z_{i+1} = z_i - \sigma_i \alpha_i, \ w_1 = P', \ z_1 = \varphi, \ w_{m+1} \approx \exp(\varphi)$$
 (12)

and

$$w_{i+1} = w_i - \sigma_i w_i 2^{-i}, \quad z_{i+1} = z_i - \sigma_i \alpha_i, \quad w_1 = P', \quad (13)$$

$$z_1 = \varphi, \ w_{m+1} \approx \exp(-\varphi), \ \varphi \in [0, \ 1.118]$$
 (14)

Here, only two iterative equations are implemented for  $w_{i+1}$  and  $z_{i+1}$ , which simplifies the structure of the device. Unfortunately, this does not lead to a reduction in the number of calculations.

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Fig. 2. Electronic implementation of the simplified CORDIC method

The end goal of our invention is to simplify the electronic implementation of the device and to reduce the number of calculation operations. First, let us consider the calculation of the function

$$x = \exp(+\varphi). \tag{15}$$

We propose dividing the input argument j into three distinct parts  $j_1, j_2, j_3$  (Fig. 3)

$$\varphi = \varphi_1 + \varphi_2 + \varphi_3 \tag{16}$$

 $\varphi = \varphi_1 + \varphi_2 + \varphi_3$  $a_{m_1}$  $a_1$  $a_2$  $a_{m_1+1}$  $a_{m_2+1}$  $a_m$ ... . . .  $a_{m_2}$ . . .  $\phi_1$  $\varphi_2$ φ<sub>3</sub> Correction unit  $x_{m_2}$  $X_{m_1}$  $X_r$ Pipelined Output LUT CORDIC multiplier Fig. 3. Hybrid CORDIC



The first part  $j_1$  occupies  $m_1$  older bits of argument j. These are fed into the LUT lookup table (Fig. 3).

$$\varphi_1 = \sum_{i=1}^{m_1} a_i 2^{-i} \tag{17}$$

The second part  $(j_2)$  of the input argument j is computed using the CORDIC method, which is implemented in the form of a single iterative equation (Fig. 4). The second part  $(j_2)$  of the input argument j occupies the subsequent  $m_2-m_1$  bits

$$\varphi_2 = \sum_{i=m_1+1}^{m_2} a_i 2^{-i} \tag{18}$$

At the end, the third calculation block provides for the multiplication of the correction angle  $j_2$ , which occupies  $m-m_2$  bits



 $\oint x_i$ Fig. 4. Proposed pipelined CORDIC

Here,  $m_1$  – the number of older bits of argument j, fed into the LUT table, which performs the function

$$x_{m_1} = P \cdot \exp(\varphi_1 + D_c) \tag{20}$$

where  $D_c$  and P – constants, which we calculate according to the formulas:

$$D_{c} = \sum_{i=m_{i}+2}^{m_{2}+1} \arctan h(2^{-i})$$
(21)

$$P = \left(\prod_{i=m_1+1}^{m_2} \sqrt{1-2^{-2i-2}}\right)^{-1}$$
(22)



The values of  $x_{m_2}$  which we read from the LUT outputs, contain *m* bits. The minimum value of  $m_1$  is obtained from the condition:

$$m_{1\min} = \left\lceil \frac{m - 10 - 2 \cdot \log_2 3}{6} \right\rceil.$$
(23)

This quantity of older bits  $m_{1_{\min}}$  gives us the ability to maintain the accuracy of calculations in *m* bits. The upper limit of  $m_1$  is limited by the value of  $m_2$  and depends solely on the LUT memory space.

The bits of the  $\varphi_1$  part, that is bits with numbers  $m_1 + 1 \dots m_2$ , are processed according to the CORDIC method. The value of  $m_2$  is chosen based on the condition

$$m_2 = \left\lceil \frac{m}{2} \right\rceil \tag{24}$$

In practice, the CORDIC equations

$$b_i = 2 \cdot a_i - 1 \tag{25}$$

and

$$x_i = x_{i-1} + b_i \cdot x_{i-1} \cdot 2^{-i-1}$$
(26)

are implemented as follows:

if 
$$a_i=1$$
, then  $b_i=1$  and  $x_i=x_{i-1}+x_{i-1}\cdot 2^{-i-1}$ , (27)

if 
$$a_i=0$$
, then  $b_i=-1$  and  $x_i=x_{i-1}-x_{i-1}\cdot 2^{-i-1}$ . (28)

It follows that CORDIC contains only one iterative equation (28). The CORDIC output gives us the code  $x_{m_2}$ . In the final stage, we use the values of *z* corrected for the  $D_{\nu}$  component of  $\varphi_{z}$ , which is calculated according to the formula

$$z = \varphi_3 + D_\nu, \qquad (29)$$

where

$$D_{\nu} = \left(\sum_{i=m_{1+1}}^{m_{3}} a_{i} \left[2^{-i} - 2 \cdot \arctan h(2^{-i-1})\right]\right)$$
(30)

$$m_3 = \left\lceil \frac{m - 5 - \log_2 3}{3} \right\rceil. \tag{31}$$

$$x_r = x_{m_2} + z \cdot x_{m_2} \,. \tag{32}$$

A patented device can be used to calculate activation functions in neural networks – this gives the possibility for a significant reduction in the amount of computation in the network teaching process. There are three types of activation functions in which the exponential function is used. These include functions



logistic 
$$f_{active}(x) = \frac{1}{1 - e^{-x}}$$
, (33)

hyperbolic 
$$f_{active}(x) = \frac{e^{x} - e^{-x}}{e^{x} + e^{-x}}$$
, (34)

exponential 
$$f_{active}(x) = e^{-x}$$
, (35)

softmax 
$$f_{active}(x) = \frac{e^x}{\sum_i e^{x_i}}$$
. (36)

## 3. Conclusions

The characteristic feature of the patented device is that the argument j of function  $y=\exp(\pm j)$  is divided into three parts and only the second part  $j_2$  is calculated using the CORDIC iteration algorithm. This solution reduces the number of computations by several times and as a result, increases the speed of the device for calculating the function  $y=\exp(\pm j)$ .

The CORDIC algorithm is implemented in the proposed device in the form of only one iterative equation. This leads to the simplification of the electronic version of this algorithm and thus reduces the cost of the device.

The time pause in the patented device will be several times shorter than the currently known solution [10]. For example, for m=32 and  $m_1=4$ , the time pause in the patented device has 14 steps, and the current solution requires 34 steps. For m=32 and  $m_1=8$ , we have 10 and 34 steps, respectively.

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